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Dynamic Properties of Angular Vertical Comb-Drive Scanning Micromirrors with Electrothermally Controlled Variable Offset

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Abstract—This paper presents the investigation of a variable angular vertical comb-drive (AVC) actuated MEMS scanning micromirror, focusing on the influence of the initial comb offset on the dynamic scan characteristics. Continuous control of the fixed comb initial vertical offset is achieved using an electrothermal actuator integrated in the AVC. Electro-mechanical simulations and experiments of the AVC structure show good agreement, with the initial vertical offset of the fixed comb reducing theoretically from 10.8µm (0mW to actuator) to 6.0µm (500mW to actuator) and experimentally from 10.4µm (0mW to actuator) to 4.9µm (510mW to actuator). Experimentally, a change of the initial vertical comb-offset from 10.4µm to 4.9µm leads to a reduction of the measured dynamic total optical scan angle (TOSA) from 27° to 19° for 100V AC actuation. Additionally, a mechanically assisted deflection of the AVC actuators to achieve an almost in-plane comb configuration, further reduces the measured dynamic TOSA to about 3°. The overall device behavior is modeled using a hybrid simulation approach combining FEM analysis of the AVC capacitance with an analytical solution of the motion equation of the scanner. The analysis shows good agreement with the experimentally measured characteristics, overall showing an increased TOSA with higher initial comb offsets.

Index Terms—Angular vertical comb-drive (AVC) actuators, scanning micromirrors, electrothermal microactuators, electrostatic force actuators

I. INTRODUCTION

ELECTROSTATIC comb-drive actuated silicon MEMS micromirrors [1] have been the subject of extensive and detailed research investigations which have led to the application of these devices in optical systems including active corner cube reflectors [2], variable optical attenuators [3, 4], wavelength multiplexers/demultiplexers [5], tunable lasers [6, 7], optical switches [8], as well as micro-projectors and imaging systems [9-13]. An example optical switch system includes an array of 1xN comb-drive actuated micromirrors in combination with a grating spectrometer to allow wavelength selective switching [8], while imaging and projection systems have been shown, for example, by using in-plane comb-drives creating resonant tilt movements [9] or in-plane rotational movements in combination with a diffraction grating [10]. Imaging and projection systems have also been proposed by using vertical offset comb-drives for actuation of cascaded 1D scanning micromirrors [11] or 2D scanning micromirrors [12, 13]. Asymmetric comb-drive actuators are a subset of general comb-drive actuators. They can generate high torques and are often used for generating stable and wide continuous range of dynamic angular motion, such as required in high scan angle micromirrors. Asymmetric comb-drive actuators are usually classified into two categories, namely staggered vertical comb-drives (SVC) and angular vertical comb-drives (AVC). The SVC category actuators have a parallel offset or have a comb zero cross-over angle between the fixed and moving combs, while the AVC category actuators have an initial non-zero cross-over angle between the fixed and moving combs.

The fabrication of SVCs in general requires complex multi mask etch steps to create the different comb thickness, either through selective etching [14, 15] or by adding isolation layers between two device layers to create the vertical offset between the active comb electrodes [16]. A further fabrication possibility has been demonstrated by using the mechanical displacement of a hinged actuator [13]. For this type of actuator a frequency tuning scheme has also been shown [17] by mechanically increasing the torsional spring stiffness using a chevron actuator. SVC actuators have also been achieved by using bulk micro-machining of a stack of silicon wafers requiring precise alignment between the individual wafers [18].

For AVC actuators various fabrication schemes have been demonstrated including the use of stiction pads [19], the reflow of photoresist [11, 12, 20], plastic deformation [21], mechanical pre-tilt of the torsion beams [22], or the use of non-actuated bimorph layers on the moving [23] or fixed comb structures [24]. In these cases the emphasis has been mainly on investigating the static behavior of the MEMS devices incorporating the AVC actuators and the influence of the comb-drive geometry and initial vertical offsets on the electro-mechanical characteristics (such as static rotation angles, pull-in stabilities or comb-drive capacitances) of the devices when operated at DC voltages. This paper presents an investigation of the dynamic behavior of AVC actuators at different comb
cross-over angles. Our AVC actuators were realized by exploiting non-uniform stresses, which exist in the silicon device layer of devices fabricated using the SOIMUMPs process [25]. These stresses, combined with thin film stresses arising from the deposition of thin gold layers, generate an out-of-plane curvature in cantilever type MEMS structures. This effect has been used advantageously in our work to design AVC actuators with curved stators and rotors, thereby generating a non-zero cross-over angle. Furthermore, we have designed novel AVC actuators whose cross-over angle can be varied by application of an electrical voltage to an electrothermal actuator integrated with the stator of the AVC. We have experimentally and theoretically investigated the dynamic response of such a variable angle AVC by incorporating it in a scanning micromirror, which enables the use of a laser detection system in the experimental characterization. To the best of our knowledge, a detailed investigation of the dynamic response of an AVC actuator for different cross-over angles has not been previously reported.

Theoretical analyses of vertical offset comb-drives have mainly concentrated on the vertical and angular displacements generated when driven by DC voltages. The analytical methods have included electrical field calculations [26] or calculation of the differential capacitance of the comb-drives [23], in both cases ignoring the electrical fringe fields. While for SVC actuators this approach might be valid, the low area overlap of the combs compared to the comb side area of AVCs requires the consideration of fringe fields. This has been achieved in [11, 12, 20] by use of a hybrid model approach, calculating the differential capacitance through an FEM simulation in combination with an analytical approach to the static torque balance. In their analysis of AVC micromirrors Hong and Sym [2] used a parametric fit of the differential capacitance to match experimentally obtained results without investigating the influence of the initial AVC angle on the dynamic performance. In the work presented here we also use a hybrid approach with a combination of FEM simulation and analytical calculations to obtain the differential capacitance. This has been done for different initial AVC cross-over angles combined with an analytical solution to the dynamic motion equation of the micromirror.

In section II the design and fabrication steps of the AVC scanning micromirror and the novel variable cross-over angle AVC comb actuator are presented. This is followed in section III by the simulation and characterization of the electrothermal actuator used to change the AVC cross-over angle. Section IV presents the experimental characterization of our AVC driven micromirror with varying initial cross-over angle of the stator/rotor combs. The hybrid theoretical model of the scanning micromirror is presented in section V with a comparison between the experimental and theoretical results in section VI and a concluding discussion in section VII.

II. DEVICE DESIGN AND FABRICATION

A. Scanning micromirror design

The layout of the single crystal silicon (SCS) MEMS scanning micromirror used to investigate the influence of the initial AVC angle on the resonant behavior of the scanner is shown in Fig. 1. The 700µm diameter mirror surface is connected via a 25µm wide circular frame to two rectangular torsion beams, each of length 745µm and width 18µm. Similar scanning micromirror designs from our group have shown maximum optical scan angles of around 80º [7]. The circular frame was chosen to reduce the dynamic mirror distortion and induce a small magnification of the mirror tilt angle in respect to the tilt introduced on the torsion beams [27]. This is achieved by combination with a second set of torsion springs of length 35µm and width 10µm connecting the frame and mirror surface in line with the main torsion springs. The moving combs of the AVC actuator are connected to a single side of one of the torsion beams, while the fixed combs are attached to a comb support beam and anchored through four electrothermal bimorph actuators (see Fig. 1). In total, 13 moving and 14 fixed combs are used to create the comb actuator with each comb finger having a length of 160µm and width of 10µm. The lateral gap between the comb electrodes is set to be 6µm, which is achieved with high accuracy as both comb electrode structures are defined in the same 10µm thick device layer, as are all other moving structures of the device. Only a single side AVC is used in this device as a further intended use is as an array configuration where generation of a small footprint limits the available space for actuators.

B. Electrothermal variable AVC actuator

The initial vertical offset of the AVC actuator is defined as the value of \( z_1 \) and \( z_2 \) (see Fig. 4) when the mirror is at its rest position. The offset is generated through stresses in the overall anchor structure of the fixed combs, consisting of the comb support beam, bimorph actuators and anchor pads. A non-uniform stress through the thickness of the silicon device layer creates an out-of-plane curvature of this single side clamped anchor structure. To increase this out-of-plane curvature a...
bimorph layer structure is created on part of the anchor structure by depositing a 1.1µm thick gold coating on its top surface during the microfabrication. These coated bimorph actuators are 250µm long and 88µm wide with two of them connected in parallel to a 90µm wide and 500µm long comb support beam and an anchor pad fixed to the silicon chip substrate (see Figs. 1 and 2).

To enable a continuous variation of the initial vertical offset, an electrothermal actuator is created using the four bimorph beams. With two adjacent bimorph actuators connected in parallel and using a separate electrical connection line, a current path is created through the bimorph anchors and the comb support beam when applying a potential difference between anchor pad 1 and 2 shown in Fig. 1. This current generates Joule heating throughout the current path and the resulting temperature change will change the curvature of the bimorph actuators. Using the material combination of SCS and gold the increase of temperature will lead to a reduction of the curvature and therefore a reduction of the initial vertical offset, $z_1$ and $z_2$. The electrothermally actuator movement range is limited by the maximum temperature in the gold layer to avoid thermal damage, as the analysis of section III will show.

C. Fabrication process

The device is fabricated using the commercially available multi-user silicon on insulator process (SOIMUMPs) offered by MEMSCAP Inc [28]. A schematic of the cross-section of the structure is shown in the inset of Fig. 1. All parts of the scanning mirror are defined in the 10µm thick phosphorus doped SCS device layer which is subjected to a single deep reactive ion etch (DRIE) step. The device layer itself is electrically isolated from the 400µm thick SCS substrate by a 1µm thick silicon-dioxide layer. A backside DRIE step is then performed through the thickness of the substrate to release the device. The gold coating layer on the electrothermal actuator beams is produced by a combination of a 500nm thick e-beam evaporated gold layer and a 600nm thick shadow mask gold layer, one deposited before and one after the DRIE process steps. A non-uniform stress in the silicon device layer generates an out-of-plane curvature of single side anchored beams, with the free end of a beam being vertically offset above the device layer [25]. This through thickness stress-gradient originates from a non-uniform doping profile through the thickness of the silicon device layer. It also leads to an intrinsic curvature of the mirror surface with an experimentally determined radius of curvature of ≈0.1m [29]. This radius of curvature is outside the usually specified limit of 0.5m for mirrors used in imaging or projection systems. However, the desired application of the scanning mirror used in this work is focused on its use as a MEMS based, active intra-cavity device in solid-state laser systems, where optically stable laser cavities can be achieved with the measured mirror surface curvature [7]. An SEM image of the fabricated device is shown in Fig. 2, with a magnified image of the comb-drive in the inset showing the initial vertical offset after fabrication.
from 0V to 3.5V was defined across the faces at the GND and +VDC ends of the model (Fig. 3). The simulation software then calculated the temperature profile in the actuator due to Joule heating, heat convection and radiation, and the coupling between thermal and mechanical deformation. The emissivity of the silicon and gold surfaces were taken to be 0.6 and 0.02 respectively. The air convection coefficient for the surfaces of the comb-drives model was taken to be 150Wm⁻²K⁻¹. Other material properties of the silicon and gold layers used for the actuator can be found in previously published work [30]. The intrinsic stress gradient of the silicon layer was 2.4MPa/µm and the residual stress within the gold layer was taken to be 300MPa. Both effects lead to the out-of-plane curvature of the fixed combs of the comb actuator, while only the intrinsic stress gradient leads to the curvature of the moving combs of the actuator.

The simulated temperature on the surface of the actuator with an applied DC voltage of 3.5V, leading to an applied DC power of 600mW, is shown in Fig. 3. The average initial vertical offset $z_1$ of the tips of the comb-drives (see inset of Fig. 4) changed from 10.8µm to 5.0µm when the applied power increased from 0mW to 600mW. The simulated change of the initial vertical offsets $z_1$ and $z_2$ for varying applied DC power is shown in Fig. 4.

### B. Experimental characterization of the variable initial vertical offset AVC

To evaluate the performance of the fabricated variable initial vertical offset AVC, the height profile of the structure was measured using a Veeco NT1100 optical white light interferometer profiler. The non-electro-thermally actuated set of combs (i.e. the rotor) of the microscanner were used as the initial height reference, with measurements being taken of the height difference at the end of the fixed combs ($z_1$, Fig. 4) and the end of the moving combs ($z_2$, Fig. 4). The fixed comb support beam (see Fig. 3) showed a maximum height difference of 0.5µm between the middle and the end of the support due to the residual stress in the silicon device layer. A DC voltage was applied between the anchor pads 1 and 2 of the fixed combs, with an electrical power level between 0mW and 510mW leading to a downward bending of the stator comb tips due to the bimorph effect between the gold and silicon layers of the stator. The fixed comb support beam uniformity changes only minimally during this actuation, showing a maximum height difference change of 0.1µm which did not significantly influence the pre-tilt angle of the AVC.

The measured average initial vertical offsets $z_1$ and $z_2$ are shown in Fig. 4, with the experimentally obtained trend matching the simulated FEM results. The maximum initial vertical offset $z_1$ of 10.4µm is measured without applying an electrical DC power, i.e. this is the value of $z_1$ of the device as fabricated. With electrothermal actuation of the bimorph stator, the initial vertical offset reduced to 4.9µm for an applied electrical power of 510mW. The DC power of the actuator was limited to 510mW to avoid thermal damage to the gold coatings of the actuator.

### IV. DYNAMIC SCANNER BEHAVIOR WITH VARYING INITIAL VERTICAL OFFSETS OF THE AVC

The experimental characterization of the influence of the initial vertical offsets $z_1$ and $z_2$ on the resonant movement behavior of the scanning micromirror was performed using two methods to vary the initial vertical offset, with the combined experimental setup shown in Fig. 5. The first method used a DC voltage control to vary the initial vertical offset as described in Section III. The second method to vary the initial vertical offset was by mechanical displacement of the fixed comb anchor structure using the sharp tip of a needle mounted on a vertical micrometer translation stage. The second displacement method was used to achieve an as close as possible in-plane configuration for the comb-drive and additionally a “negative” initial vertical offset, with the position of the stator combs lying below the plane of the surrounding silicon die, and therefore below the initial position of the rotor combs. A mechanical displacement had to be used for this part as the damage limit of the electrothermal actuator prevented a displacement range leading to an in-plane comb-drive configuration. The resulting total optical scan angle (TOSA) for each initial vertical offset of the device was
measured using the reflection of a probe HeNe laser beam on the mirror which was projected onto a screen. Simultaneously the initial vertical offsets were measured using a Veeco NT1100 optical white light interferometer profiler.

A. Electrically varied initial vertical offset

The initial vertical offset was electrically varied while the microscanner was driven at its resonance frequency. A DC power varying from 0mW to 510mW was applied between the anchor pads 1 and 2 of the stator combs to vary the initial vertical offset as previously explained. The microscanner dynamic displacement was generated using an AC waveform from a signal generator (Agilent 33250A) which was amplified by a 20x voltage amplifier (FLC A400) before being applied to the scanner. The AC actuation voltages consisted of an offset square-wave waveform with peak-peak value of 80V and 100V at a frequency of around 6 kHz, corresponding to the torsional resonance mode of the scanner. TOSA measurements were taken at four initial vertical offset positions, starting from the maximum initial vertical offset $z_1$ of the fixed comb tip of 10.4µm to the lowest possible initial vertical offset of 4.9µm which ensured that the electrothermal actuator was not thermally damaged.

The resulting frequency response curves showing the measured TOSA are presented in Fig. 6. For both values of AC voltages applied to the scanner a reduction of the maximum TOSA is observed for lower values of initial vertical offset. Additionally a resonance frequency increase of 5Hz (around 0.08%) can be seen as the initial vertical offset is reduced from its maximum value to its minimum value. For 80V AC actuation a TOSA of 21º at 6.005kHz was measured for the non-actuated initial vertical offset (i.e. without electrothermal actuation) while application of 510mW to the fixed comb anchor pads led to a TOSA of 14º at 6.010kHz. For the 100V AC actuation a similar behavior was observed with TOSAs of 27º at 6.010kHz and 19º at 6.015kHz respectively for the two initial vertical offset values.

B. Mechanically varied initial vertical offset

The mechanical variation of the initial vertical offset is conducted on the same microscanner as the electrical offset control, creating the vertical displacement of the comb anchors by using a needle tip. The needle is fixed on a vertical micrometer stage and mounted at an angle to allow the simultaneous imaging of the fixed comb displacement using the optical profiler while adjusting the displacement of the combs. The needle tip has a width of around 100µm and is carefully centered between the two sets of bimorph beams to exert a balanced downward pressure leading to a change in initial vertical offset, whilst making great effort to minimize any twisting of the comb structures. Four displacement cases were chosen: (i) initial vertical offset at the fixed comb tip at the rest position of 10.4µm (ii) about half way between the rest position and an in-plane alignment relating to 3.9µm (iii) as close as possible in-plane configuration with an initial vertical offset of 0.5µm and (iv) a negative initial vertical offset of -1.7µm placing the fixed combs beneath the moving comb rest position. Due to the tip width being smaller than the overall comb-anchor width, the mechanical displacement of the combs leads to a slight non-homogeneous comb tip displacement with higher displacements in the middle compared to the edges of the anchor showing a maximum difference of 1.2µm for the in-plane configuration. The quoted offset values are therefore average offsets.

Fig. 6. Frequency response curve of AVC scanning micromirror with electrical variation of the initial comb-offset. (a) Actuation of the scanner with a square-wave voltage signal with $V=80V$, (b) actuation of the scanner with a square-wave voltage signal with $V=100V$. 
The frequency response curves for the four cases are shown in Fig. 7 for square-wave AC actuation voltages of 80V and 120V peak-peak. Similar to the electrically induced initial vertical offset change, a higher initial vertical offset leads to higher TOSA values at the resonance peak. For 80V actuation the TOSA reduces from 21º at the maximum initial vertical offset to about 2º for the lowest initial vertical offset difference and rises again to 9º when reaching the negative initial vertical offset. As already mentioned, no perfect in-plane comb alignment could be reached with the mechanical displacement, leading to an average initial vertical offset of 0.5µm at the tip of the fixed combs (z₁) and -0.9µm at the tip of the moving combs (z₂). This initial vertical offset in both vertical directions plays a part in the low TOSA frequency response for this case. For the 120V actuation a similar behavior is observed with the maximum TOSA reducing from 32º for the initial vertical offset z₁ to about 3º for the closest in-plane configuration and rising to 17º when moving to the negative initial vertical offset. The initial vertical offsets for both actuation voltages are identical since frequency sweeps have been carried out for both without changing the mechanical setup.

Comparing experimental results obtained from both initial vertical offset control techniques shows the same behavior, namely lower absolute values of initial vertical offsets leading to a lower angular response of the microscanner. At the same time a small frequency shift in the region of 5Hz towards higher frequencies can be seen at the resonance peaks when moving from the highest initial vertical offsets to the lowest achieved initial vertical offsets.

V. HYBRID THEORETICAL MODEL FOR AVC SCANNER

To investigate the expected dynamic behavior of the microscanner for varying initial vertical offsets a theoretical model is used, combining an FEM simulation of the capacitance between the comb fingers of the AVC actuator with the numerical solution of the basic equation of motion of the scanner for torsional displacements. The model for the numerical solution is based on the model used by Hong and Symns [2] for the evaluation of a fixed initial vertical offset AVC. The angular displacement $\theta$ of the scanning mirror around its torsion axis is governed by the equation of motion of a forced oscillator and is written as

$$I\ddot{\theta} + 2\xi\omega_0\dot{\theta} + \omega_0^2\theta = \frac{1}{2} V^2(t) \frac{\partial C(\theta,t)}{\partial \theta}.$$  \hspace{1cm} (1)

where $I$ is the torsional moment of inertia of the scanner, $\omega_0$ is the angular resonant frequency, $\xi$ the damping coefficient, $V(t)$ the applied voltage driving signal and $C(\theta,t)$ the angular and time dependent capacitance of the AVC actuator. The damping coefficient is related to the experimentally measured Q-factor of the resonance by

$$\xi = \frac{1}{2Q} = \frac{f_2 - f_1}{2f_0}.$$  \hspace{1cm} (2)

where $f_0$ is the resonance frequency and $f_2$ and $f_1$ are the frequency values where the angular response has dropped to a value of $1/\sqrt{2}$ times the maximum angle. The spring related part of the motion equation can be written in form of a torsional spring with spring constant $k_0$ where

$$k_0 = \omega_0^2 I.$$  \hspace{1cm} (3)

If a nonlinear behavior of the torsional spring is present then this spring constant has to be adapted to include higher order terms, with the major influence being a cubic nonlinearity term for symmetric displacement movements as is present in our microscanner [31]. The modified spring constant then takes the form

$$k_\theta = k_0 \left(1 + k_2 \theta^2 \right).$$  \hspace{1cm} (4)
where $k_1$ is the resulting spring constant and $k_2$ is the cubic nonlinear spring constant. Positive $k_2$ values lead to spring hardening behavior while negative $k_2$ values show a spring softening effect \[32\]. The right hand side of equation (1) comprises the electrostatic torque applied through the AVC actuator. The electrical actuation voltage can be written as

$$V^2(t) = \frac{1}{4} V_0^2 \left[ l + \text{sgn} \left( \sin \left( 2\pi ft \right) \right) \right]^2, \quad (5)$$

with $V_0$ being the amplitude of the square-wave voltage signal and $f$ the device actuation frequency. To account for the variable capacitance distributions for varying initial vertical offsets and for the fringe fields between the comb electrodes, which are not negligible for AVC actuators \[20\], a 2D FEM simulation is used combined with an analytical expression for the instantaneous relative comb positions between the rotor and stator.

A. Electrostatic fringe field analysis

To determine the 2D electrical field distribution for a varying height difference $z$ of a cross section of the fixed and moving combs in a unit cell of the comb drive a FEM simulation using the software COMSOL Multiphysics is performed. A unit cell consists of the end-on cross section of a half width of a moving and a half width of a fixed comb as shown in Fig. 8. The use of this unit cell reduces the computational requirements, with the use of only a half comb per unit cell to allow the summing of multiple unit cells to calculate the overall comb-drive electrical field distribution \[20\]. The simulation solves the electrical potential distribution of the unit cell for an applied voltage $V_{\text{FEM}}$ between both half combs. This is done by defining both half combs as terminals in the FEM software with their respective material properties set to the used SCS. The 6µm gap between the combs is filled with air by using the COMSOL in-built material properties. The electrical energy density $w_e$ of the unit cell is then calculated using a surface integration of the simulated potential distribution over the 2D area using the FEM software. This process is repeated for varying height steps $z$ between the two half combs, leading to a distribution of $w_e$ over varying height differences. The electrical field distribution for three example cross-section offsets of 0µm, 5µm and 10µm is shown in Fig. 8 (with $V_{\text{FEM}}=80\text{V}$), also showing the necessity to include the fringe field in the capacitance determination as major parts of the field distribution are outside the overlap area.

B. Capacitance evaluation over angular movement range

To determine the capacitance $C(\theta)$ of the actuator for a specific mirror tilt angle $\theta$, with inclusion of the contributions of the fringe fields, the electrical energy density of the 2D unit cell of section A above needs to be integrated over the length of a comb-finger while taking the instantaneous vertical comb offset $z(l,\theta)$ for each section of the comb fingers into account. The vertical comb offset distribution along the length of the comb-overlap can be formulated using the definitions in Fig. 9. When including the tilt angle $\theta$ of the moving comb it can be written as

$$z(l,\theta) = l \cdot \sin(\theta - \theta_i) + z_i + l_2 \cdot \sin(\theta). \quad (6)$$

$l$ is the length variable originating from the tip of the fixed comb, $\theta_i$ is the initial angle between the moving and fixed comb, $z_i$ is the initial vertical offset at the tip of the fixed comb and $l_2$ is the length along the moving comb between the rotation axis and the start of the comb overlap on the fixed comb tip. $\theta_i$ and $z_i$ are inputs which are experimentally obtained from the optical profiler measurements of the fabricated comb-drives. From this expression a calculation of the overall actuator capacitance for a specific tilt angle of the moving combs is achieved using...
N is the number of gaps between the moving and fixed combs and \( l_1 \) the length of the comb overlap, as can be seen in Fig. 9. This integration is undertaken using Matlab with a fit to the electrical energy density distribution and the measured initial comb-offset values as inputs. The differential capacitance \( \partial C / \partial \theta \) is then approximated using

\[
\frac{\partial C}{\partial \theta} = C(p) - C(p - 1) \\
\frac{\theta(p) - \theta(p - 1)}{\theta(p) - \theta(p - 1)}
\]

with \( p \) being the running variable through the calculated set of capacitance values for varying angles \( \theta \). An example distribution of the actuator capacitance and differential capacitance for an initial vertical offset \( z_1 \) of 10.4µm is shown in Fig. 10. The considered angular range is in this case an optical scan angle of ±45º.

C. Analytical motion equation evaluation

To use the above determined differential capacitance distribution in the torque related term of the motion equation, a 20th order Fourier fit is used on the dataset. A numerical solution of the ordinary differential equation in (1) is constructed using Matlab with its in-built Runge-Kutta solver in the function ODE45. To solve the second order differential equation it needs to be split into two first order equations in the form

\[
\dot{\theta}_1 = \theta_2 \\
\dot{\theta}_2 = -4\pi \xi f_0 \theta_2 - \frac{k_0}{I} \theta_1 + \frac{1}{2I} V^2(t) \frac{\partial C}{\partial \theta_1}
\]

To determine the moment of inertia \( I \) and the spring constant \( k_0 \) a FEM modal analysis of the scanner layout is used (as seen in Fig. 11), resulting in a value for the moment of inertia of \( I = 3.963 \times 10^{-16} \text{ kg} \cdot \text{m}^2 \) and a resonance frequency of \( f_0 = 6255 \text{ Hz} \). This simulated resonance frequency overestimates the experimentally observed frequency of the fabricated device by 250Hz. Therefore the frequency used as simulation input is based on the experimental results with an initial vertical offset \( z_1 \) of 10.4µm, through which the resonance frequency can be extrapolated as \( f_0 = 5990 \text{ Hz} \), rather than using the FEM value of 6255Hz. The only other parameter relying on prior measurements is the damping coefficient \( \xi \), which is determined through the Q-factor of the frequency response curve when the thermal actuator is not actuated. This experimental estimation is necessary as an accurate analytical calculation of the damping behavior for the micromirror structure is outside the scope of this work. To calculate the frequency response curve, multiple iterations of the ODE45 function are required with varying actuation frequency inputs \( f \). The achievable TOSA for each frequency step is extracted after the simulated oscillation has reached its settling time.

Within the simulation model the influence of the frame and its second pair of torsion beams connecting the mirror surface is neglected and no coupled system of differential equations is constructed. This is justified due to a low amplification factor between the angular displacement of the frame and the mirror which was measured to be below 1.2. The mirror and frame combination can therefore be modeled as a stiff structure in a first order approximation.

VI. Verification of Simulation Results with Experimental Data

The simulation results obtained in the previous sections are verified using the experimental data obtained for the 80V AC actuation. Four initial vertical offsets were used, two \( z_1 \) values of 10.4µm and 4.9µm for comparison with the electrothermally varied initial vertical offsets, and two \( z_1 \) values of 0.5µm and -1.7µm for comparison with the mechanically obtained initial vertical offsets. A summary of

![Fig. 10. Calculated capacitance and differential capacitance for an initial AVC offset of 10µm between the fixed and moving combs.](image1.jpg)

![Fig. 11. FEM simulation of torsion resonance mode shape.](image2.jpg)
the input parameters for the simulation is presented in Table 1, with the spring constant $k_0$ being calculated through the modal FEM results for $I$ and $f_0$. The two initial vertical offset parameters $z_1$ and $z_2$ for each offset case are the average values taken from the optical profiler measurements and are used for determining the differential capacitance distribution. The damping coefficient is taken from the 10.4µm initial vertical offset frequency response curve of the electrical displacement case in Fig. 6(a) with a Q-factor of 166. The actuation voltage $V_o$ used in the simulation model is 56V. A cubic nonlinear spring constant of $k_2=0.33$ was included in the modeling as the experimental measurements of the TOSAs showed no nonlinear characteristics (e.g. bending of the frequency response curve) while an initial simulation showed a nonlinear response at movement angles above TOSA=15º. This resulted from electrostatic field induced nonlinearities in the simulation, which were counteracted by modifying the simulation to include the nonlinear behavior of the torsional spring.

The resulting simulated frequency response curves for the four actuation cases are shown as lines in Fig. 12, with the open points depicting the experimentally obtained values from section IV. The simulation results show a good agreement with the experimental values with the trend of lower TOSA for lower initial vertical offsets being also present in the simulation results. The experimentally observed frequency shift for lower initial vertical offsets is also shown by the simulation results, with the simulation having a shift of 20Hz between the initial vertical offset of 10.4µm and the closest in-plane configuration with 0.5µm and therefore introducing an even higher shift than experimentally observed.

VII. CONCLUSION

Both the experimental and simulation results have shown higher dynamic scan angles of the AVC actuated scanning micromirrors for higher initial vertical offsets. The AVC scanning micromirrors are themselves novel in design and operation, since the relative position of the stator and rotor combs can be changed by electrical control. Therefore, the comb-offset for the AVC can be adjusted precisely. No perfect in-plane configuration can be achieved between the stator and rotor combs; however, a low initial fixed comb angle of 0.5º is the closest experimental realization of an in-plane configuration. The full set of possible initial vertical offsets for our AVC microscanners were investigated, with the maximum possible initial vertical offset limited in order to prevent thermal damage of the gold layer coating the electrothermal actuator. Alternative designs for the fixed comb anchor actuator may increase the possible initial vertical offset variation range and could lead to the determination of a maximum initial vertical offset range after which the dynamic scanner behavior will not have a further increase in the angular frequency response. The hybrid simulation model presented shows a good agreement with the experimental data with an already mentioned overestimation of the frequency shift for changing maximum TOSAs. Only a cubic nonlinearity is used for modeling the spring constant with a possible higher agreement between the experimental and simulated data when considering further orders of nonlinearity. These would require an optimization for a fit of the simulations to the experimental data, which was not conducted in the model developed as apart from the damping coefficient, all other model parameters were based on theoretical values.

Overall, an investigation of the dynamic behavior of AVC actuated microscanners has been presented, with the focus on studying the impact of the initial vertical offset of the actuators on the dynamic scan angles generated. Similar to previous investigations of the static scan angles, a higher initial vertical offset was shown to give a higher angular dynamic response at resonance. Experimental results matched the theoretical findings of a hybrid simulation model.

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![Fig. 12. Comparison of the simulated resonance curves and the measurement results for 80V actuation. 10µm and 5µm offset measurements are from the electrical offset actuation and 0µm and -2µm offset measurements are from the mechanical offset actuation.](image-url)
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